A Factorization Method Approach to the Biharmonic Transmission Problem in Absorbing Media

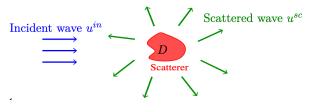
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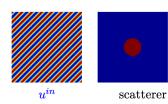
Graduate Research Day November 15, 2025



Direct & Inverse Scattering



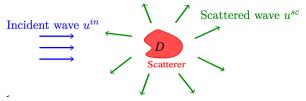
 $\begin{cases} u \text{ governed by a partial differential equation} \\ u = u^{\text{in}} + u^{\text{sc}} \quad \text{(total wave)} \\ u^{\text{sc}} \text{ is an outgoing wave} \quad \text{(radiation condition)} \end{cases}$



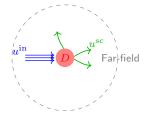




Direct & Inverse Scattering



- Direct Problem: Given incident wave uⁱⁿ and scatterer D, find the scattered wave u^{sc}.
- **Inverse Problem:** Given the (far-field) of scattered wave u^{sc} , find the scatterer D.



Setting: Biharmonic Transmission Problem

 \blacktriangleright Out-of-plane displacement $u=u^{\rm in}+u^{\rm sc}$ satisfies the biharmonic transmission equation

$$\begin{split} \left(\Delta^2 - \kappa^4 n(x)\right) u(x) &= 0 \quad \text{in } \mathbb{R}^2, \quad n \in L^\infty(\mathbb{R}^2), \text{ } \mathrm{supp}(n-1) = \overline{D}. \\ u^{\mathrm{in}}(x,d;\kappa) &= \exp\left(i\kappa x \cdot d\right), \quad d \in \mathbb{S}^1, \quad \kappa > 0 : \text{ } \mathrm{wavenumber} \end{split}$$

▶ Transmission conditions on ∂D :

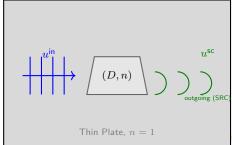
$$[\![u]\!] = 0, \qquad [\![\partial_{\nu} u]\!] = 0,$$

 $[\![\Delta u]\!] = 0, \qquad [\![\partial_{\nu} \Delta u]\!] = 0.$

where $[\![\varphi]\!] = \varphi_+|_{\partial D} - \varphi_-|_{\partial D}$ denotes the jump across the interface.

 $ightharpoonup u^{\rm sc}$ and $\Delta u^{\rm sc}$ satisfy a suitable radiation condition at infinity (SRC).

Rafael Ayala²



Absorbing Scatterer Condition and Well-Posedness

Direct Problem: Given the incident field u^i , find the scattered field u^s for a known penetrable medium (n,D):

$$(\Delta^2 - \kappa^4 n(x))\, u = 0 \text{ in } \mathbb{R}^2, \quad u = u^{\mathrm{in}} + u^{\mathrm{sc}}, \quad u^{\mathrm{sc}} \,\&\, \Delta u^{\mathrm{sc}} \text{ satisfies SRC}.$$

- ▶ Well-posedness (Fredholm of index zero) is established via a variational approach.
- ▶ Uniqueness: In the biharmonic case, requires the absorbing scatterer condition

$$\Im(n(x)) \geq \alpha > 0 \quad \text{a.e. } x \in D$$

which is not necessary in the acoustic case.

Theorem (Ceja-Ayala, R., Harris, I. & Sánchez-Vizuet, T., 2025)

For a penetrable medium (n,D) satisfying $\Im(n) \ge \alpha > 0$ a.e. in D and a given incident field $u^{\rm in}$:

- \blacktriangleright The direct scattering problem has a unique solution u^{sc} .
- ► The solution operator is Fredholm of index zero.
- \blacktriangleright The scattered field u^{sc} satisfies the estimate

$$||u^{\rm sc}||_{H^2(D)} \le C||u^{\rm in}||_{L^2(D)}$$



Lippmann-Schwinger Type Representation

 $\mbox{\bf Goal:}$ Express the $\mbox{\bf unique}$ scattered field solution $u^{\rm sc}$ in terms of volume integral

Idea: Apply the Green's theorem for x in both the *interior* and *exterior* of the scatterer D. (Note χ_D is characteristic function over D)

Lippmann-Schwinger Type Representation

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Idea: Apply the Green's theorem for x in both the *interior* and *exterior* of the scatterer D. (Note χ_D is characteristic function over D)

Interior $(x \in D)$

$$u^{\mathrm{sc}}(x)\chi_D = \int_D G(x,z;\kappa)[\kappa^4 u^{\mathrm{sc}}(z) - \Delta^2 u^{\mathrm{sc}}(z)] \,\mathrm{d}z + \mathrm{boundary\ terms\ on\ }\partial D.$$

Exterior $(x \notin D)$

$$u^{\text{sc}}(x)(1-\chi_D) = \text{boundary terms on } \partial D \cup \partial B_R.$$

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Combine both: adding the interior and exterior representations gives

Lippmann-Schwinger (L-S) Type Equation

$$u^{\mathsf{sc}}(x,d) = \kappa^4 \int_D (n(z) - 1) G(x,z;\kappa) \left[u^{\mathsf{in}}(z,d) + u^{\mathsf{sc}}(z,d) \right] \mathrm{d}z.$$

Interpretation:

► $G(x,z;\kappa)$ is the Green's function of biharmonic operator $(\Delta^2 - \kappa^4)$.

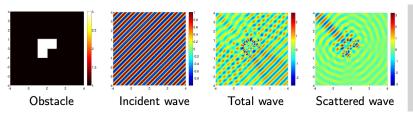
The Inverse Problem

Far-field intuition: Far from the scatterer, the scattered wave behaves like an outgoing cylindrical wave. Its amplitude and phase in each observation direction define the far-field pattern.

$$u^{\mathrm{sc}}(x,d) = \frac{e^{i\pi/4}}{\sqrt{8\pi\kappa}} \frac{e^{i\kappa|x|}}{\sqrt{|x|}} \frac{u^{\infty}(\hat{x},d;\kappa) + O(|x|^{-3/2}), \quad |x| \to \infty, \ \hat{x} = \frac{x}{|x|}.$$

- $u^{\infty}(\hat{x}, d; \kappa)$: measured far-field pattern in direction of observation \hat{x} for incident waves $d \in \mathbb{S}^1$
- ▶ Inverse problem: Recover the 'absorbing' penetrable scatterer characterized by the pair $(n, D = \operatorname{supp}(n-1))$ its shape, size, and location given the full far-field data

$$\{u^{\infty}(\hat{x},d;\kappa): \hat{x} \in \mathbb{S}^1, d \in \mathbb{S}^1\}.$$



Adapted from: A. Lechleiter, Making the Invisible Visible: Imaging Techniques for Inverse Problems (2009)

Rafael Ayala²



We now define the (compact!) far-field operator as $F:L^2(\mathbb{S}^1) o L^2(\mathbb{S}^1)$

$$(\mathbf{F}g)(\hat{x}) = \int_{\mathbb{S}^1} \mathbf{u}^{\infty}(\hat{x}, \mathbf{d}; \kappa) g(\mathbf{d}) \, ds(\mathbf{d}).$$

 ${\it Fg}$ is far field corresponding to the incident field (Herglotz wave function)

$$v_g(x) = \int_{\mathbb{S}^1} e^{i\kappa x \cdot d} g(d) \, ds(d)$$

Note: v_g refers to superposition of plane waves.

Inverse Problem

The **inverse problem** also reads: given F find D = supp(n-1)!

How to reconstruct *D* from *F*? Nonlinear optimization, local linearization, domain decomposition... time-consuming!, Requires forward solver!

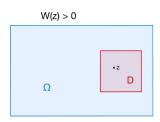


Qualitative/Sampling Methods

Class of Reconstruction Methods

Qualitative Methods: Construct a binary criterium that decides whether a grid point z belongs to the support of the region in a computational simple manner.

- ► Advantages: Requires little a priori information on the scatterer. Very fast, no forward solver needed, fully direct method.
- ▶ Disadvantages: requires a lot of data, can only recover qualitative information.





The Range Test Property

Define the Herglotz wave operator

$$(Hg)(x)=v_g(x),\quad x\in D,\quad L^2(\mathbb{S}^1)\to L^2(D)\quad \text{bounded}$$

and its adjoint

$$(\boldsymbol{H}^*\varphi)(\hat{x}) = \int_D e^{-i\kappa \hat{x}\cdot \boldsymbol{y}} \; \varphi(\boldsymbol{y}) \, d\boldsymbol{y}, \quad \hat{x} \in \mathbb{S}^1, \quad L^2(D) \to L^2(\mathbb{S}^1) \quad \text{bounded}$$

For a sampling point $z \in \mathbb{R}^2$, consider the range test equation

$$H^*\varphi = \phi_z$$
, where $\phi_z(\hat{x}) = e^{-i\kappa \hat{x}\cdot z}$.

Note: ϕ_z is a **test function** and far-field pattern for the fundamental solution of the Helmholtz equation

Range Test Property

Let $D = \operatorname{supp}(n-1)$ be a bounded open set with connected complement. Then,

$$\phi_z \in \text{Range}(H^*) \iff z \in D.$$

Key. Proof via a unique continuation argument.

 H^* is a **bounded operator** depending on the (unknown) domain D.

Key Question

Can we relate $\operatorname{Range}(H^*)$, which depends on the unknown D, to the known far-field operator F?



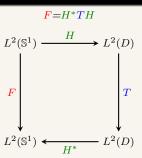
Factorization Method Works by Factorization!

- ▶ Herglotz wave operator: $H: L^2(\mathbb{S}^1) \to L^2(D), Hg = v_q|_D$
- ▶ Define $T: L^2(D) \to L^2(D)$ by

$$Tf = \frac{\kappa^2}{2}(n-1)(f+w(f)|_D)$$

where $(\Delta^2 - \kappa^4 n(x))w = \kappa^4 (n-1)f$ in $\mathbb{R}^2 + (SRCs)$

heorem



Task: Relate the Range(H^*) to the known operator F.

Factorization is From Lippmann-Schwinger

Starting point:

$$w(x) = \kappa^4 \int_D (n(y) - 1) G(x, y) [w(y) + f(y)] dy.$$

Using the asymptotic form of G(x,y) as $|x| \to \infty$:

$$G(x,y) \sim \frac{e^{i\kappa r}}{\sqrt{r}} \underbrace{\frac{\gamma}{2\kappa^2} \, e^{-i\kappa \hat{x} \cdot y}}_{\text{far-field term}}, \quad \gamma = \frac{e^{i\pi/4}}{\sqrt{8\pi\kappa}}$$

Then the far-field pattern becomes:

$$w^{\infty}(\hat{x}) = \frac{\kappa^2}{2} \int_D e^{-i\kappa \hat{x} \cdot y} (n-1) \left[f(y) + w(y) \right] dy.$$

Hence, for a Herglotz incident field f = Hg,

$$(\mathbf{F}g)(\hat{x}) = \frac{\kappa^2}{2} \int_D e^{-i\kappa \hat{x} \cdot y} (n-1) \left[Hg(y) + w(y) \right] \mathrm{d}y.$$

Recognizing the operator structure:

$$F = H^* T H, \qquad Tf = \frac{\kappa^2}{2} (n-1)(f+w).$$

Key Property of the Middle Operator T

Operator factorization:

$$F = H^* T H, \qquad Tf = \frac{\kappa^2}{2} (n-1) (f+w)$$

Define the imaginary parts:

$$\Im(F) := \frac{F - F^*}{2i}, \qquad \Im(T) := \frac{T - T^*}{2i},$$

which are self-adjoint & positive.

$$\Im(F) = H^* \Im(T) H$$

We have: Range($(\Im(F))^{1/2}$)=Range(H^*), which follows from the following result.

Theorem ((Ceja-Ayala, R., Harris, I. & Ozochiawaeze, G. (2025))

Let the scatterer D = supp(n-1) be absorbing, i.e.,

$$\Im(n(x)) \ge \alpha > 0$$
 a.e. $x \in D$.

Then the middle operator $\Im(T)$ is coercive on $L^2(D)$, that is, $\exists \mu > 0$ so that

$$(\Im(T)f, f)_{L^{2}(D)} = \Im(Tf, f)_{L^{2}(D)} \ge \mu ||f||_{L^{2}(D)}^{2}$$

for all $f \in L^2(D)$.

Key. Proof by contradiction, exploit biharmonic wave decomposition for w obtained from factoring $(\Delta^2 - \kappa^4)w = (\Delta + \kappa^2)(\Delta - \kappa)w$

Reconstruction of D by the FM

We have the positive definite, compact operator

$$\Im(F) := \frac{F - F^*}{2i} : L^2(\mathbb{S}^1) \to L^2(\mathbb{S}^1).$$

We also have

$$\mathsf{Range}((\Im(F))^{1/2}) = \mathsf{Range}(H^*)$$

so

$$z \in D \iff \phi_z \in \mathsf{Range}((\Im(F))^{1/2})$$

The Factorization Method (FM)

Let the pair $(\lambda_j, \psi_j) \in \mathbb{R}^+ \times L^2(\mathbb{S}^1)$ be the orthonormal eigensystem of $\Im(F)$. Then by the Picard range criterion we have:

$$z \in D \iff W(z) = \left[\sum_{j=1}^{\infty} \frac{\left| (\phi_z, \psi_j)_{L^2(\mathbb{S}^1)} \right|^2}{\lambda_j} \right]^{-1} > 0.$$

Thus, we have complete characterization of penetrable scatterer D:

$$\chi_D(z) = \mathrm{sgn}(W(z)) \coloneqq \begin{cases} 1, & x \in D \\ 0, & x \not\in D \end{cases}$$



Born Approximation for Weak Scatterers

Integral operator:

$$(Kf)(x) := \int_D G(x,y) (n(y) - 1) f(y) dy$$

L-S Type Equation then becomes:

$$u^s = Ku^{\text{in}} + Ku^{\text{sc}} \iff u^s = (I - K)^{-1}Ku^{\text{in}}$$

Born approximation:

$$u^{\rm sc} pprox \underbrace{Ku^{\rm in}}_{\rm first-order}$$

Far-field (linearized):

$$u^{\infty}(\hat{x}, d) \approx \frac{\kappa^2}{2} \int_D (n - 1) e^{-i\kappa \hat{x} \cdot y} u^{\text{in}}(y) dy$$
$$= \frac{\kappa^2}{2} \int_D (n - 1) e^{-i\kappa \hat{x} \cdot y} e^{i\kappa d \cdot y} dy$$
$$= \kappa^2 / 2 \cdot \widehat{(n - 1)}(\xi), \quad \xi \coloneqq \kappa (d - \hat{x})$$

Key intuition: small/weak scatterer \Rightarrow far-field \approx Fourier transform of contrast.



Comparing Exact Far-Field & Born Approximation

Assume a small disk scatterer $D=B_\epsilon$, $0<\epsilon\le 1$, with constant material parameter n=1.0+1.5i. The direct scattering problem reads:

$$\Delta^2 u^{\mathrm{sc}} - \kappa^4 u^{\mathrm{sc}} = 0 \quad \text{in } \mathbb{R}^2 \setminus \overline{B_\epsilon}, \quad \Delta^2 u - \kappa^4 u = 0 \quad \text{in } B_\epsilon,$$

with boundary (transmission) conditions at $r = \epsilon$:

$$u^{\mathrm{sc}} - u = -u^{\mathrm{in}}, \quad \partial_r u^{\mathrm{sc}} - \partial_r u = -\partial_r u^{\mathrm{in}}, \quad \Delta u^{\mathrm{sc}} - \Delta u = -\Delta u^{\mathrm{in}}, \quad \partial_r \Delta u^{\mathrm{sc}} - \partial_r \Delta u = -\partial_r \Delta u^{\mathrm{in}}.$$

Expand incident plane wave via Jacobi-Anger:

$$u^{\text{in}}(r,\theta) = \sum_{\ell=0}^{\infty} i^{\ell} J_{\ell}(\kappa r) e^{i\ell(\theta-\phi)}, \quad d = (\cos\phi, \sin\phi)$$

Separation of variables and the boundary conditions at $r=\epsilon$ give a linear system:

$$\mathbb{M}\mathbf{u} = \mathbf{f}, \quad \mathbf{u} = [a_{\ell}, b_{\ell}, c_{\ell}, d_{\ell}]^{T}, \quad \mathbf{f} = [-J_{\ell}(\kappa \epsilon), -\kappa J'_{\ell}(\kappa \epsilon), \kappa^{2} J_{\ell}(\kappa \epsilon), \kappa^{3} J'_{\ell}(\kappa \epsilon)]^{T}.$$

Once solved, the exact biharmonic far-field pattern is

$$u^{\infty}(\theta,\phi) = \frac{4}{\mathrm{i}} \sum_{\ell=0}^{\infty} a_{\ell} \, e^{i\ell(\theta-\phi)}.$$



Computing the Matrix M for a Disk Scatterer

Fourier Series Ansatz for the Scattered Field.

We decompose u^{sc} and u into propagative $((\Delta + \kappa^2)$ and evanescent $(\Delta - \kappa^2)$ components:

$$u_{\mathsf{H}}(r,\theta) = \sum_{|\ell|=0}^{\infty} i^{\ell} a_{\ell} H_{\ell}^{(1)}(\kappa r) e^{i\ell(\theta-\phi)}, \quad u_{\mathsf{M}}(r,\theta) = \sum_{|\ell|=0}^{\infty} i^{\ell} b_{\ell} H_{\ell}^{(1)}(i\kappa r) e^{i\ell(\theta-\phi)}.$$

Inside the Scatterer.

The total field $u = u_{pr} + u_{ev}$ is expanded as

$$u_{\mathrm{pr}}(r,\theta) = \sum_{|\ell|=0}^{\infty} i^{\ell} c_{\ell} J_{\ell}(\kappa n^{1/4} r) e^{i\ell(\theta-\phi)}, \quad u_{\mathrm{ev}}(r,\theta) = \sum_{|\ell|=0}^{\infty} i^{\ell} d_{\ell} J_{\ell}(i\kappa n^{1/4} r) e^{i\ell(\theta-\phi)}.$$

Key Observation.

$$\Delta u_{\rm pr} = -\kappa^2 \sqrt{n} \, u_{\rm pr}, \quad \Delta u_{\rm ev} = \kappa^2 \sqrt{n} \, u_{\rm ev}.$$

Linear System from Boundary Conditions.

At $r = \epsilon$, the four boundary conditions give

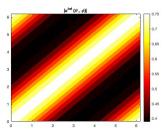
$$M\mathbf{u} = \mathbf{f}$$

which determines the coefficients $a_{\ell}, b_{\ell}, c_{\ell}, d_{\ell}$.

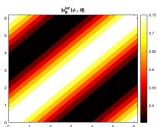


Biharmonic Far-Field Pattern vs Born Approximation ($\varepsilon=0.5$)

Exact Far-Field Pattern



Born Approximation



$$\begin{split} \mathbf{F}_{\text{exact}} &= [u^{\infty}(\theta_i, \phi_j)]_{i,j=1}^{64}, \quad \mathbf{F}_{\text{Born}} = [u_{\text{B}}^{\infty}(\theta_i, \phi_j)]_{i,j=1}^{64}, \\ \hat{x} &= (\cos\theta, \sin\theta), \quad d = (\cos\phi, \sin\phi), \quad n = 1.0 + 1.5i \end{split}$$

Absolute Error $\|\mathbf{F}_{\mathsf{exact}} - \mathbf{F}_{\mathsf{Born}}\|_{\infty}$ for Small Disks B_{ϵ}

| ϵ | Error |
|------------|--------|
| 1.00 | 0.6480 |
| 0.90 | 0.5008 |
| 0.80 | 0.3784 |
| 0.70 | 0.2774 |
| 0.60 | 0.1935 |
| 0.50 | 0.2134 |

FM Discretization and Imaging Functional

Algorithm 1 Factorization Method

Input: Wavenumber κ , will fix regularization $\alpha=10^{-5}$, noise level δ .

Step 1: Sample Directions.

Compute 64 equally spaced angles $\theta_i = 2\pi(i-1)/64$, $\hat{x}_i = d_i = (\cos\theta_i, \sin\theta_i)$.

Step 2: Assemble Far-Field Matrix.

Evaluate $\mathbf{F}(i,j) = u^{\infty}(\hat{x}_i, d_j)$ and form $\mathbf{F} \in \mathbb{C}^{64 \times 64}$.

Step 3: Add Noise (optional).

 $\mathbf{F}^{\delta}(i,j) = \mathbf{F}(i,j)(1+\delta \mathbf{R}(i,j)), \quad \mathbf{R} \in \mathbb{C}^{64 \times 64} (\text{error matrix})$

Step 4: Compute SVD of Imaginary Part.

 $\Im(\mathbf{F}) = \frac{\mathbf{F} - \mathbf{F}^*}{2i} = \sum_j \sigma_j \, \mathbf{u}_j \mathbf{v}_j^*.$

Step 5: FM Indicator.

$$W_{\mathrm{FM}}(z) = \left[\sum_{j} \frac{\phi^{2}(\sigma_{j}; \alpha)}{\sigma_{j}} |(\mathbf{u}_{j}, \boldsymbol{\ell}_{z})|^{2} \right]^{-1}, \quad \boldsymbol{\ell}_{z} = [e^{-i\kappa\hat{x}_{i} \cdot z}].$$

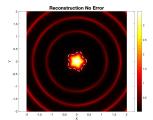
Tikhonov Filter Function

Tikhonov regularization: $\phi(t;\alpha) = \frac{t^2}{t^2 + \alpha}, \quad \alpha = 10^{-5}$

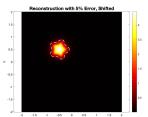


FM Reconstruction: Star-shaped Scatterer (Penetrable, n = 2.5 + 1.5i)

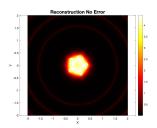




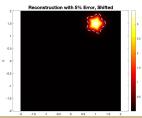
(5% noise, shifted (-0.5, 0.5))



 $k=3\pi$ (no noise)



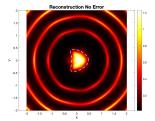
(5% noise, shifted (1, 1.5)



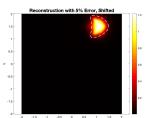


FM Reconstruction: Kite-shaped Scatterer (Penetrable, n = 2.5 + 1.5i)

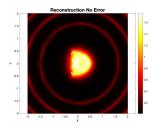




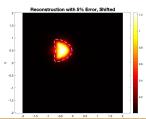
5% noise, shifted (-0.5, 0.5)



 $k=3\pi$ (no noise)

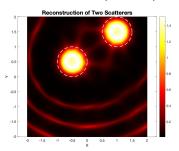


5% noise, shifted (1, 1.5)

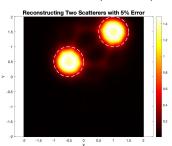


FM Reconstructions: Disk Scatterers (Penetrable, $\kappa=2\pi$, n=2.5+1.5i)

Reconstruction (No Noise)



Reconstruction (5% Noise)



FM reconstructions of a small disk scatterer $D = B_{1/2}$.

Conclusion

- ► Extended the Factorization Method (FM) to biharmonic scattering in penetrable, absorbing media.
- Provided numerical reconstructions showing FM's effectiveness for small scatterers and limited/noisy data.
- verified heuristically Born approximation valid first order linearization for weak scatterers.
- Open problems / future directions:
 - FM for biharmonic scattering by impenetrable cavities is still open.
 - Extending reconstruction methods to near-field biharmonic scattering for a penetrable medium remains open.
- Overall, FM provides a qualitative reconstruction framework beyond classical acoustic, elastic, & electromagnetic scattering, with extensions to complex media feasible.

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